An Improved Speed and Torque Performance of ANFIS based Direct Torque Controlled Induction Motor Drive

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Abstract—Unlike vector control of induction motor direct Torque Control (DTC) has several advantages. This paper presents intelligent control scheme with ANFIS controller (adaptive neuro fuzzy inference system) to adjust the speed of the direct torque control of induction motor drive. The performance of conventional DTC in Induction Motor drives consisting of PI controller suffers from complicated tuning and Overshoot problems. This control scheme uses the speed error calculated from reference speed and estimated speed which generates the estimated Torque and compared with the actual Torque. Simulation studies have been carried out for different operating conditions of the drive system; the results have been presented and compared with those of the conventional method. From the simulation results it can be observed that when there is a step change in the load torque, the momentary decrease in speed with the proposed method is less. Finally the proposed method will provide less THD in steady state current ripple when compared to the conventional DTC algorithm.

Index Terms— Induction motor, direct torque control, PI, ANFIS and THD.

1 INTRODUCTION

In 1971, F. Blaschke presented the first paper on the field-oriented control (FOC) for induction motors. In recent years, commercial applications of vector controlled induction motor drives have greatly increased. The disadvantage of this control scheme is inclusion of the pulse encoder, PWM modulator and the indirect torque control [1].

The above disadvantages are overcome by a direct torque control (DTC) technique. In this technique both flux and torque are controlled by using a hysteresis controller [2]. In this method the PWM modulator is replaced by an optimal switching logic and it results in a good torque response which is better than that of the vector control technique without any mechanical transducers [3]. Main disadvantage of conventional DTC technique is having very high torque ripple. In recent years several techniques have been developed to reduce the torque ripple [4]. A new DTC scheme based on discrete space vector modulation (DSVM) technique describes the control system, which generates more number of voltage vectors than used in conventional DTC scheme. In this method an increased number of voltage vectors allow a smooth variation in torque and flux, without increasing the complexity of conventional DTC. The switching tables are generated from the analysis of the equations associating the applied voltage vector to the corresponding torque and flux variations. Industrial applications exhibit significant non-linearities, so that the performance may deteriorate if conventional controller such as PI controller is used [5-6]. For these reasons, it is necessary to develop a suitable controller which is capable of handling linear as well as non linear systems. Fuzzy controllers are suitable to control the above aspects. Hence, fuzzy logic control systems combining conventional control techniques leads to more effective control design with improved system performance and robustness [7-10].

The main objective of this paper is to develop ANFIS based algorithm uses sophisticated switching tables to generate the PWM signals to the inverter for the control of an induction motor drive. Consequently speed, torque, flux and angle are estimated by using the stator voltages and stator currents in the adaptive motor model. AnANFIS based speed controller is designed and incorporated in this model to improve the speed performance. Several numerical simulations have been presented to emphasize the drive performance based on the proposed technique for various loading conditions, at different speeds and during the speed reversals.

2 CONVENTIONAL DTC PRINCIPLES

The electro magnetic torque in the stationary reference frame is given by

\[ \tau = \frac{3}{2} p (\lambda_d i_q - \lambda_q i_d) \]

Fig.1 Stator and rotor flux-linkage space vectors
\[
T_e = \frac{3 \rho}{2} \frac{L_{im}}{\sigma L_s L_r} \left| \lambda_s \right| \left| \lambda_r \right| \sin \eta \quad (1)
\]

\[
\sigma = \text{Leakage coefficient} = I - \left( \frac{L_{im}^2}{L_s L_r} \right) \quad (2)
\]

Where \( \eta \) = angle between the stator and rotor flux linkage space vectors, as shown in Fig.1.

From equation (1) it is clear that the motor torque can be varied by changing the rotor or stator flux vectors. The rotor time constant of a standard squirrel-cage induction machine is very large, thus the rotor flux linkage changes slowly compared to the stator flux linkage. However, during a short transient, the rotor flux is almost unchanged. Thus rapid changes of the electromagnetic torque can be produced by rotating the stator flux in the required direction, which is determined by the torque command. On the other hand the stator flux can instantaneously be accelerated or decelerated by applying proper stator voltage phasors. Depending on the position of the stator flux, it is possible to switch on the suitable voltage vectors to control both flux and torque. By using the torque and flux errors, an optimum switching table is constructed for picking up suitable voltage vectors to increase or decrease torque and flux, so that the torque and flux linkage errors are to be controlled within the hysteresis band. The block diagram for the conventional DTC is shown in fig.2.

\[\text{Fig. 3 Structure of ANFIS controller}\]

Layer 1: (Input Layer)
Input layer represents input variables of controller, they are actual speed and reference speed respectively. This layer just supplies the input values \( x_i \) to the next layer, where \( i = 1 \) to \( n \)

Layer 2: (Fuzzification Layer)
This layer receives the input values from the first layer it creates membership function for the respective input variables and these are inputs to the next layer

Layer 3: (Rule layer)
Each node (each neuron) in this layer performs the precondition matching of the fuzzy rules, i.e., they compute the activation level of each rule, the number of layers being equal to the number of fuzzy rules. Each node of these layers calculates the correction which are normalized.

Layer 4: (Defuzzification Layer)
It provides the output values “\( y \)” resulting from the inference of rules. Connections between the layers 1 & 4 are weighted by the fuzzy rules that represent another set of parameters for the neuro fuzzy network.

Layer 5: (Output Layer)
In this layer all the inputs coming from the layer 4 sums up and transforms the fuzzy classification results into a crisp values.

\[\text{Fig.2 Block diagram of conventional DTC}\]

3 ANFIS CONTROLLER

ANFIS is based on fuzzy inference system and this system uses the given input and output data to build fuzzy inference system. First a training data set that contains the desired input/output data pairs of target systems to be modeled is required. The design parameters required for any ANFIS controller are number of data pairs, training data sets and checking data sets. For training the number of epochs to be chosen to start the training, learning results to be verified after mentioning the step size. Then the designed ANFIS has two inputs namely, the actual motor speed and reference speed while the output is the torque, which is used to generate current. Structure of ANFIS speed controller is shown in Fig. 3 and it is based on the five-layer feed forward fuzzy neural network.
4 ANFIS BASED DIRECT TORQUE CONTROLLED INDUCTION MOTOR DRIVE:

The block diagram of proposed algorithm is as shown in Fig. 4. As in conventional direct torque control, the proposed algorithm generates d-axis and q-axis reference stator currents, which are at synchronously rotating reference frame. Here we are using two - level hysteresis controller and lookup table. Thus, the proposed algorithm eliminates time consuming PWM procedure. The generated d- and q- axis current commands are compared with their actual current values obtained from the measured phase currents with ANFIS controller.

The current errors are used to produce d- and q-axes flags as inputs to the switching table. A third input to the table determines the sector through which the current is passing. Based on the outputs of hysteresis controllers and position of the stator current, the optimum switching table will be constructed. This gives the optimum selection of the switching voltage space vectors for all the possible stator current vector positions.

Fig. 4 Block diagram of ANFIS based direct torque controlled induction motor drive

5 SIMULATION RESULTS AND DISCUSSIONS

To validate the proposed algorithms, numerical simulation studies have been carried out by using Matlab-Simulink. For the simulation studies the dc link voltage is taken as 540V. The parameters of the induction motor used in this paper are \( R_s = 1.57 \text{ohm}, R_r = 1.21 \text{ohm}, L_m = 0.165 \text{H}, L_s = 0.17 \text{H}, L_r = 0.17 \text{H} \) and \( J = 0.089 \text{Kg-m}^2 \). The simulation results are shown from Fig. 5 – Fig. 14.

Fig. 5 starting transients with conventional DTC induction motor
Fig. 6 steady state plots with conventional DTC induction motor drive

Fig. 7 Transients in speed, torque, and currents during step change in load (a load torque of 25 N·m is applied at 0.7 s and removed at 0.9 s) with conventional DTC induction motor drive

Fig. 8 Performance during speed reversal operation (from +1000 rpm to -1000 rpm) with conventional DTC induction motor drive

Fig. 9 Harmonic spectra of steady state line current with conventional DTC induction motor drive

Fig. 10 Starting transients with ANFIS based DTC induction motor drive

Fig. 11 steady state plots with ANFIS based DTC induction motor drive
6  CONCLUSION

The proposed ANFIS based induction motor drive is providing better performance when compared to conventional algorithm. The proposed controller is evaluated under simulations for a variety of operating conditions of the drive system and results demonstrate the effectiveness of these control structures to improve the performance of the drive system. It uses the instantaneous errors in d-and q axes stator currents and sector information to select the suitable voltage vector. Hence, the proposed algorithm uses a predetermined switching table instead of a much more time consuming PWM procedure in conventional FOC algorithm. From the simulation results it can be observed that when there is a step change in the load torque, the momentary decrease in speed with the proposed method is less. Finally the proposed method will provide less THD in steady state current ripple when compared to the conventional DTC algorithm.

REFERENCES